Enhancing LTE with Cloud-RAN and Load-Controlled Parasitic Antenna Arrays


ABSTRACT

Cloud radio access network systems, consisting of remote radio heads densely distributed in a coverage area and connected by optical fibers to a cloud infrastructure with large computational capabilities, have the potential to meet the ambitious objectives of next generation mobile networks. Actual implementations of C-RANs tackle fundamental technical and economic challenges. In this article, we present an end-to-end solution for practically implementable C-RANs by providing innovative solutions to key issues such as the design of cost-effective hardware and power-effective signals for RRHs, efficient design and distribution of data and control traffic for coordinated communications, and conception of a flexible and elastic architecture supporting dynamic allocation of both the densely distributed RRHs and the centralized processing resources in the cloud to create virtual base stations. More specifically, we propose a novel antenna array architecture called load-controlled parasitic antenna array (LCPAA) where multiple antennas are fed by a single RF chain. Energy- and spectral-efficient modulation as well as signaling schemes that are easy to implement are also provided. Additionally, the design presented for the fronthaul enables flexibility and elasticity in resource allocation to support BS virtualization. A layered design of information control for the proposed end-to-end solution is presented. The feasibility and effectiveness of such an LCPAA-enabled C-RAN system setup has been validated through an over-the-air demonstration.

INTRODUCTION

Mobile data traffic is expected to continuously rise in the future [1]. In view of the radio spectrum’s scarcity, it is necessary to incorporate in LTE standard technologies and techniques that have the potential to enhance the capacity of next-generation LTE-based systems. We present a solution based on the synergy of universal frequency reuse, cell densification, and coordinated multipoint (CoMP) to improve spectral efficiency (SE), increase system capacity, and mitigate inter-cell interference (ICI). Albeit promising, these techniques present implementation challenges: the current network architecture struggles to meet the stringent inter-cell communication throughput, and there are also the latency requirements of the various CoMP “flavors,” thus reducing the effectiveness. Moreover, while it is well known that typically the multi-user spatial multiplexing (MU-SM) and interference management capabilities of CoMP benefit from more antennas installed at the base station, in practice conventional antenna arrays make use of a limited number of elements. This is because of cost and power constraints due to the standard practice of feeding each antenna by a separate RF module. Additionally, this is also because of the reduction of the radiation efficiency observed when the antennas are too closely spaced, due to the resulting strong electromagnetic coupling. This limits the performance gains of CoMP techniques.

The system described in this work brings together two elements that address the aforementioned issues: cloud radio access network (C-RAN) architecture [1] and load-controlled parasitic antenna arrays (LCPAAs). The former envisages that the baseband units (BBUs) of the BSs are centralized, virtualized via cloud technology, and connected to the remote radio heads (RRHs) — located at the cell sites — through an optical aggregation network called fronthaul (FH). C-RAN ensures the timely communication between cooperating cells and accommodates the heavy data/control traffic load related to CoMP [1]. It provides flexibility and allows for cost-effective deployment of dense networks due to the reduced number and energy needs of the cell site facilities. Additionally, C-RAN results in efficient utilization of the available resources, because of the dynamic allocation of BBUs and RRHs to “virtual BSs,” according to the traffic and radio coverage demands. LCPAAs, on the other hand, use fewer RF chains than antenna elements and exploit the mutual coupling among the antennas to shape the radiation pattern. This provides a high number of antenna array degrees of freedom (DoF), thus boosting the performance of the system, while reducing the hardware cost, complexity, energy consumption, and footprint.

Matteo Artuso and Henrik L. Christiansen are with the Technical University of Denmark; Dora Boviz, Aravinthan Gopalasingham, and Laurent Roulet are with Nokia Bell Labs; Aleksandra Checko is with the Technical University of Denmark and MTI Radiocomp; Bruno Clerckx and Borzoo Rassouli are with Imperial College London; Laura Cottatellucci, David Gesbert, and Haifan Yin are with EURUCOM; Bobby Gizas, Dimitrios Ntaikos, Konstantinos Ntougias, and Constantinos B. Papadias are with AIT; Faheem Khan, Tharmalingam Ratnarajah, and Lin Zhou are with the University of Edinburgh; Jean-Marc Kelif and Stephane Senecal are with Orange; Ralf Müller and Mohammad Ali Sedaghat are with the Norwegian University of Science and Technology.
We have to note that while the C-RAN paradigm has been extensively studied in the past few years as a CoMP and (network) densification enabler, the proposed setup, which considers RRHs equipped with LCPAAs, presents clear advantages over conventional configurations. That is, the use of LCPAAs leads to an additional beamforming (BF) gain for a given number of RF chains or cost reduction for a target BF gain.

Since its introduction in 2010, C-RAN has received tremendous interest from academia and industry, and there are multiple works addressing topics of optimal resource allocation, RRH and virtual BBU clustering, fronthaul optimization, and enhancing C-RAN with software defined networking (SDN) as well as presenting testbeds and prototypes. Reconfigurable solutions based on C-RAN to improve resource utilization are proposed in [2, references therein].

The FH evolution has received considerable attention recently, and will lead to different functional splits for different deployment scenarios, most likely resulting in a packet-based transport network. To enable this, the following standardization activities are in place:

- IEEE 1914, “Next Generation Fronthaul Interface,” aims to define architectures for fronthaul transport, requirements and definitions for the FH network, as well as possible functional split schemes.
- IEEE 802.1Qbu, “Bridges and Bridged Networks—Amendment: Enhancements for Scheduled Traffic,” standardizes source scheduling in order to reduce jitter of FH streams in Ethernet switches by ensuring that, at specific times, only one traffic stream has access to the network.
- IEEE 802.1Qbu, “Bridges and Bridged Networks—Amendment: Frame Preemption,” standardizes preemption, which can suspend the transmission of lower-priority traffic when FH traffic needs to be transmitted.
- IEEE 802.1CM, “Time-Sensitive Networking for Fronthaul,” is a profile that selects features, options, configurations, defaults, protocols, and procedures — defined in the IEEE 802 standard — for switches and stations for the FH use case.

Moreover, the EU project iCirrus — among others3 — also investigates using Ethernet for FH networks. Reference [3] identifies benefits and challenges of Ethernet-based FH (EFH).

The novelty of this work lies in combining LCPAAs with software-defined C-RAN and with EFH, as well as presenting an end-to-end (E2E) prototype. Thus, cost and performance benefits of C-RAN can be further enhanced by usage of LCPAA:

1. There are cost and energy savings due to the utilization of a reduced number of RF chains for given antenna array DoFs in comparison with conventional antenna systems.
2. There are reduced complexity and simpler control circuits for performing beam shaping/steering, in contrast to phased antenna arrays.
3. Also, there are the benefits of lighter weight and more compact size, due to the closer vicinity of antenna elements, leading to arrays that fit into the RRH’s limited volume as well as EFH:

1. Optimal support of decreased FH data rate: Variable bit rate data from the new functional split can be transported in packets.
2. Flexibility in deployments and operation: Traffic aggregation and switching is supported, allowing optimal resource allocation and maximization of multiplexing gain.
3. Cost optimization: EFH networks could utilize widely deployed existing Ethernet deployments so that in-phase/quadrature (I/Q) data shares resources with other types of traffic.

Our contributions on the EFH are aligned with the aforementioned standardization efforts. The considered solution does not consist of simply using diverse technological components and transmission methods; it also introduces enhancements of the employed CoMP techniques, network protocols, and technologies (Fig. 1). Moreover, it defines design and implementation guidelines that span every aspect of the system architecture, from the air interface to the EFH network and from virtualization to E2E network design. The considered solution is complementary to the existing state of the art of C-RAN. The aim of this article is to present the system architecture, describe the aforementioned advances, study the fundamental performance limits for the related techniques under this setup, and propose an E2E system prototype.

**SYSTEM ARCHITECTURE**

The evolution of LTE requires rethinking of the network E2E and the synergies between the different components. This ensures a harmonious integration of advancements in different research areas. Figure 1 depicts the proposed integrated vision for future mobile communication networks.

Starting from the antennas used at the RRHs, we present a new design and control in the family of LCPAAs, based on the concept of load-modulated array (LMA), which ensures high flexibility as it contains elements tunable at the symbol level. Introducing this element raises the issue of understanding its fundamental characteristics, especially compared to other variations of LCPAA systems, such as electrically steerable passive array radiators (ESPARs). Our results on the fundamental limits of both cases point out that for the case of ESPAR, signal preprocessing at the transmitter can ensure significant improvements, while in the case of LMA, fixing the sum power of the antennas is beneficial.

As the network is equipped with this advanced control and antenna system, it allows consideration for advanced multiple-input multiple-output (MIMO) radio access (RA) schemes at the radio interface. However, the volume of data and control to be processed increases dramatically and poses serious challenges in terms of efficient and timely processing of the signal. We therefore designed a novel RA scheme that capitalizes on the synergies between ESPAR antennas and C-RAN. Additionally, we also explored the general properties of distributed RRH systems to emulate and approximate centralized joint processing under practical constraints.

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3 Related projects in the area are 5G-Crosshaul (http://5g-crosshaul.eu/, accessed 16.08.2016) and 5G-xhaul (https://5g-ppp.eu/5g-xhaul/, accessed 16.08.2016).
Finally, to ensure that this advanced processing can take place, the entire FH connecting the RRHs to the BBU pool has to be designed, keeping in mind the requirements of the aforementioned RA techniques and at the same time ensuring flexibility and elasticity. To tackle this problem, an EFH has been considered, together with a complete virtualization of the BBU functionalities in the pool. This clearly ensures the flexibility and elasticity required in future communications systems so that advanced CoMP techniques can be implemented and a multiplexing gain can be exploited. However, it introduces challenges such as synchronization of traffic, dimensioning, and the impact of the network design on the applications used by the user equipment (UE). We thus address these challenges in the overall design of the system by taking a holistic view of the system.

**PROPOSED ENHANCEMENTS FOR FUTURE C-RAN NETWORKS**

**NOVEL ANTENNA DESIGN**

We consider two types of LCPAAs: ESPARs and LMAs. In ESPARs, the parasitic antenna elements are not directly connected to the single power amplifier, but fed by current induced by the active antenna element. LMAs are the novel single-RF multiple-antenna transmitters introduced in [5]. In contrast to ESPARs, in LMAs, all the antenna elements are fed by a central power amplifier via transmission lines.

A block diagram of an LMA transmitter is presented in Fig. 2. In LMAs, the source is a simple oscillator. Each antenna element is connected to a load modulator $L_M$ tunable at symbol rate. The variations of the load are steered via a level shifter from digital baseband, and modulate the signal onto the amplified carrier. Each load modulator contains some passive and some tunable elements. Since the load modulators are steered directly from digital baseband, there is no need for either mixers or digital-to-analog converters (DACs). The signal on each antenna is set by changing the state of the tunable elements in the respective load modulator. In LMAs, the signal at the input of the central power amplifier is a constant envelope sinusoid signal. Thus, the central power amplifier may be nonlinear and efficient. If the number of antenna elements is large, the law of large numbers ensures that the aggregate load seen by the power amplifier hardly varies, little power is reflected into the circulator, and the overall architecture operates at high power efficiency. For 100 antenna elements, a power efficiency of 70 percent at error vector magnitude (EVM) of –40 dB is reported in [4].
A load modulator is a linear two-port network, such as a \( \Pi \) or \( T \) network, containing some tunable elements (e.g., variable capacitors or switched capacitor banks). A load modulator can be implemented using either softly tunable elements (e.g., varactor diodes) or hard RF-switches (e.g., pin diodes). All digital processes (e.g., mapping, coding, precoding) occur in the digital baseband block. The latter also computes the appropriate bias states for all diodes at all symbol times. The bias voltages of the diodes are generated by level shifter blocks that execute commands by the digital baseband block every symbol clock cycle.

**FUNDAMENTAL LIMITS**

Here we discuss limits and flexibility of multi-antenna architectures fed by single RF modules (ESPAR and LMA) in efficiently using frequency spectrum and energy.

**ESPAR:** In the ESPAR antenna (EA) configuration, for some signals’ transmission, the input resistance becomes negative, leading to oscillatory/unstable behavior. We showed that instead of these actual signals, one can transmit approximate signals, and we derived a simple closed-form expression that enables easy generation of the approximate signals at the transmitter [5]. The EAs utilizing this novel closed-form expressions are denoted as EAs with preprocessing (EA-P). According to simulation results, conducted for various communication scenarios, the EA-P transmitter provides significant improvement over the performance of the EA without preprocessing and performs nearly the same as the standard multi-antenna system (SMA).

Additionally, in [6] we studied the impact in terms of energy efficiency (EE) of using EAs in the RRHs of a C-RAN for both single- and multi-UE systems. The closed-form expressions were derived to obtain the EAs’ configurations to maximize the EE. We have shown that such an EA-equipped system provides better EE performance while providing similar symbol error rate (SER) performance as an SMA.

**LMAs:** In LMA, the sum power of all the antennas is generated by a central amplifier, which necessitates peak and average power constraints. The capacity of the MIMO channel under these constraints was investigated in [7], where it was found that the support of the capacity-achieving distribution is a finite set of hyper-spheres for the identity channel matrix. Interestingly, when the average power constraint is relaxed, if the number of antennas is large enough, the capacity has a closed form solution and is achieved by constant amplitude signaling at the peak power. Several upper and lower bounds for the capacity of the non-identity channel matrix are proposed.

In order to make the central power amplifier efficient, we have proposed to fix the sum power of all the antennas for all channel uses in [8]. We name this new type of modulation phase modulation on the hypersphere (PMH). PMH is suitable for UEs with few antenna elements. Figure 3 shows the mutual information of uniform PMH in an uplink multiple access channel. It is assumed that the number of antenna elements at the BS and the aggregated number of antenna elements at the UEs are equal and go to infinity. It is observed that with only a few antennas at the UEs, the mutual information becomes very close to the mutual information that results from Gaussian input. This shows that fixing the sum power does not cause significant rate loss, while it makes the power amplifier in LMA efficient.

**TRANSMISSION TECHNIQUES**

**Arbitrary Channel-Dependent Precoding Utilizing Single-RF ESPAR Antennas:** In order to perform closed-loop MIMO transmission using single-RF ESPAR antennas, we map the precoded symbols on the antenna currents [9]. Unfortunately, for some input signal constellations the required loading values may lead to circuit instability. We can overcome this issue and perform arbitrary channel-dependent precoding by splitting the problem into two parts:

- **Beamforming:** A radiation beam pattern is shaped using any valid method.
- **Precoding:** Precoding-based transmission is performed over the employed beam.
Virtual access transmission
Joint ZFBF (CoMO) single-RF

3.1. responsible beam combinations.

• A limited feedback scenario, where the system operates according to the aforementioned phases, but the UEs report their SINR to the BSs; this information used for the selection of the best beam pair (green).

• In schemes where precoding is used, the sum-rate curves do not floor since the joint ZFBF transmission nulls IUI, in contrast to the non-precoded transmission schemes.

• The ESPAR-enabled precoding-based setup outperforms the omnidirectional setup over the entire SNR range due to the power gain related to the application of transmit beamforming.

• CSI-based transmission significantly outperforms SINR-based transmission, since in the former case the beam pair selection process is ZFBF-driven, while in the latter case the procedures of beam pair selection and ZFBF precoding design are decoupled.

Additionally, Fig. 5 (right) shows the over-the-air (OTA) sum rates measured with and without precoding using the LCPAAs. The testbed setup was configured for two active users transmitting concurrently while having their arrays placed within a distance of a few meters. We connected two directional LCPAAs at each transmitter port and sent two independent spatial streams of data at the same time and in the same frequency band, and used two omnidirectional monopole antennas at the receiver ports to disentangle the received data symbols. With zero-forcing precoding we achieved quadruple the capacity.

Statistical MIMO Processing for Distributed RRHs: The processing of MIMO signals in a network with many distributed ESPAR-enabled RRHs hinges on the large volume of data to be processed. This poses problems related to computing complexity and signaling overhead as both channel state and I/Q samples have to be conveyed from the RRHs to the BBU pool. As a countermeasure, algorithms allowing for a partitioning of the signal processing across the various levels of the network can lift the signaling and complexity burden, thus making them very desirable. As such, ESPAR antennas can offer local beamforming capabilities at the individual RRH level, while keeping the number of active RF branches low. Another strategy relies on antenna dimension reduction via so-called layered precoding or decoding. The layered approach consists of exploiting an inner antenna combiner that is based on channel statistics alone, followed by an outer (reduced-dimension) combiner which is designed based on instantaneous CSI. The statistical combiner exploits the intrinsic spatial structure in the UE’s channel vectors revealed by their covariance matrices. In particular, the existence of the low rank of the covariance matrix can enable interference mitigation without requiring instantaneous CSI. In [10], this property is evidenced for massive antenna array channels, be they linear arrays or distributed arrays such as RRHs.

MIMO System with Imperfect CSIT: In multi-user MIMO (MU-MIMO) networks, knowledge of CSI at the transmitter (CSIT) is crucial for interference mitigation and beamforming purposes. However, in practice, CSIT may be imperfect due to channel estimation error and/or feedback latency. CSIT inaccuracy results in MU interference that is the primary bottleneck of MIMO wireless networks as highlighted, for example, in LTE-A CoMP studies. In ESPAR and

An example of this solution is depicted in Fig. 4. We assume a C-RAN setup with two RRHs connected to a BBU pool and two single-antenna UEs. Each RRH is equipped with a five-element single-RF ESPAR that is able to generate four predetermined beams. Hence, there are $4^2 = 16$ possible beam combinations.

The system operates in three phases:

1. Learning: The UEs estimate the gains of the cross and direct links corresponding to each beam combination and report these estimates back to their BSs.

2. Selection: The $2 \times 2$ channel matrix for each beam combination is compiled at the BBU pool for the RRHs from these estimates. Then, based on that channel state information (CSI) knowledge, the best pair of beams is jointly selected for transmission.

3. Transmission: A joint transmission (JT) scheme is chosen for the precoding. We opt for zero-forcing beamforming (ZFBF), which is applied on top of the selected beams to null the resulting inter-user interference (IUI) at each single-antenna receiver.

Figure 5 (left) illustrates the attained average sum-rate throughput, as calculated via numerical simulations for signal-to-noise ratio (SNR) values in the range $[-20, 30]$ dB (red). For comparison, three more scenarios are simulated:

- An equivalent single-RF scenario where each RRH is equipped with a conventional omnidirectional antenna, and joint ZFBF transmission takes place (blue).
- A non-precoded transmission scenario, where the best beam pair is selected based on signal-to-interference-plus-noise ratio (SINR) feedback information but no precoding takes place. In this case, there is no need for the UEs to send back the channel gains of the selected beam pair (purple).
- A limited feedback scenario, where the system operates according to the aforementioned phases, but the UEs report their SINR to the BSs; this information used for the selection of the best beam pair (green).

Based on the simulation results on the left of Fig. 5 we can point out:

- The ESPAR-enabled precoding-based setup outperforms the omnidirectional setup over the entire SNR range due to the power gain related to the application of transmit beamforming.

- CSI-based transmission significantly outperforms SINR-based transmission, since in the former case the beam pair selection process is ZFBF-driven, while in the latter case the procedures of beam pair selection and ZFBF precoding design are decoupled.

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SINR (dB) ZFBF 20 25 30

LMA-based networks, the knowledge of CSIT also plays a crucial role as in conventional arrays. A complete characterization of general MIMO networks with perfect, delayed, or unknown CSIT is an open problem. In [11], we provide some answers toward this goal for a K-user MU-MIMO downlink. An outer bound for the DoF region is derived based on the marginal probabilities of CSIT availability of each UE. A set of inequalities is proposed that captures not only the marginals, but also the joint CSIT distribution. This shows that in general, marginal probabilities are not sufficient for characterizing the DoF region. In other words, any two CSIT patterns with the same marginals but different joint statistics can have different DoF regions, and some CSIT patterns are much more suitable than others. This provides new perspective on how to design novel and efficient CSI feedback mechanisms in LTE-A.

**NETWORK AND PROTOCOLS DESIGN**

**Cooperative Signal Processing in C-RAN:** C-RAN enables CoMP at the physical layer (PHY) by allowing low-latency communication between BBU serving neighboring cells. In uplink, cell edge UE’s signal can be affected by severe interferences from co-channel UEs in congested scenarios where the same frequency is used by UEs on the edge of two adjacent cells. Since ICI signals are received by both of the two neighboring cells, thanks to centralization this interference can be exploited. We use multi-cell joint scheduling for interfering UEs, so they transmit on the same frequency resource and their signals can be detected in the BBU pool with conventional MIMO detection on the signals received by two (or more) RRHs. By slightly increasing transmit power, the same error rate as in the interference-free case is reached using two times less resources [12].

**Multiplexing Gain:** BBU centralization also enables cost savings due to statistical multiplexing gain (MG). However, they can be compromised because of the large capacity required in the FH when a “traditional” functional split based on the Common Public Radio Interface (CPRI) is used [1]. Deployment decisions thus benefit from our results focused on calculating MGs achieved in the BBU pool and on the FH with various functional splits. For a CPRI-based FH, the MG on traffic-dependent BBU resources is 1.2–1.6, and higher for 30 percent office and 70 percent residential BSs. Reference [1] further examines MGs for BBU as well as FH links for the “traditional” CPRI-based split, the user/cell split, and the packet-data convergence protocol (PDCP)/radio link control (RLC) split.

**Synchronization:** Existing Ethernet networks can be reused for C-RAN FH to leverage their wide deployment, cost efficiency, and switching possibility. With the new aforementioned user/cell functional split we expect variable bit rate data streams on the FH, so a packet-based transport will optimize resource utilization in the network. The main challenges in using EFH are ensuring low delay and clock recovery. We built a demo of SDN-controlled EFH carrying 2.5 Gb/s I/Q data streams where the delay to pass one switch was within 3 μs [1]. Moreover, we studied factors influencing synchronization in an architecture employing the IEEE 1588 protocol between a master clock in the BBU pool and a slave node in the RRH or in the CPRI-to-Ethernet gateway (GW).

**Application-Layer Performances:** Dimensioning of the EFH should reflect the maximum latency requirements of the communication protocols used. A too-long FH latency triggers the retransmission timers in the hybrid automatic retransmission request (HARQ), which leads to a waste of network capacity with a chain effect that propagates to the higher layers, thus seriously affecting the performances of the applications running in the UEs. We used network and protocol simulations to estimate how FH latency affects the page response time of typical web browsing traffic [13]. Simulation results show that up to 0.6 ms of latency can be tolerated; however, optimal results are obtained with 0.1–0.2 ms of latency, which ensure a page response time of ~3.6 s for the 80th percentile of the UEs.

**Transport Control Protocol Optimization:**

The natural following optimization for C-RAN has been focused on the Transmission Control Protocol (TCP) due to its central role in providing a reliable E2E communication link for many applications in the UEs. The aforementioned network model has been integrated with offline PHY processing and BBU signal processing delay esti-
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**DEMONSTRATOR FOR LTE EVOLUTION TECHNOLOGIES**

**OVERVIEW**

A selection of the enhancements listed above has been integrated in a prototype demonstrator. The system concept has been validated in a setup based on LCPAA and C-RAN. The diagram of the system is shown in Fig. 6, while the key elements of the demonstrator are briefly presented in the following. Two laptops are used as UEs, one connected with an LTE dongle, the other with a USRP B200 software-defined radio (SDR) card. An overall E2E prototype has been presented in [15], albeit without precoding. The prototype for this portion of the system is presented in Fig. 5, and test results discussed above.

**KEY DEMONSTRATOR ELEMENTS**

**LCPAA**: Beam steering for the LCPAAs is controlled from the BBU with each antenna making use of four parasitic elements and tuned at 2.6 GHz that can create four distinct predetermined beams with a width of 40° each. At each symbol time, the receiver feeds back to the transmitter the channel quality indicator (CQI) for each beam pair combination, and the best pair is selected. In low-mobility or static scenarios, the selection normally take place once and for all during an initial training phase.

**RRH**: The RRH used in the demonstration is a NG3 Nokia unit configured in 10 MHz bandwidth with a single-input single-output (SISO) interface to the antenna system and connected to the CPRI-to-Ethernet GW via a rate 3 CPRI link.

**CPRI-to-Ethernet GW**: Nokia’s CPRI-to-Ethernet GW converts I/Q data from CPRI to 10 Gb Ethernet (10GbE) and vice versa. Commercial RRHs typically support CPRI, so the GW is necessary to connect them to an EFH. The GW is based on the Zyq board from Xilinx, a system on chip (SOC) embedding a field programmable gate array (FPGA), and a dual core ARM processor. The FPGA carries the I/Q from Ethernet to CPRI with minimum delay, and the ARM processor is used for configuration and monitoring as well as configuration and management functions (C&M). C&M messages from the CPRI links and non-I/Q packets from the 10GbE links are routed to the ARM processor. Two GWs are used, each with one 10GbE link to the EFH network and one rate 3 CPRI link to the RRH port.

**EFH**: The EFH is built using three Nokia enterprise Omniswitch 6900s, ensuring 10 Gb/s latency on the order of sub-microseconds. They also support the OpenFlow protocol; thus, the EFH is controlled by an SDN controller. This allows routing and dynamic mapping of flows between the GW and the virtualized BBUs, depending on network conditions and load in each BBU. Ethernet packets are sent from the BBU through the switches to the GW; here they are re-encapsulated to CPRI toward a legacy RRH and vice versa.

**Virtualised BBU Pool**: OpenStack is used for the virtualized BBU pool with OAI virtual eNodeBs within Docker containers. Docker is a virtualization technology for deploying applications and processes inside Linux containers. This ensures higher efficiency compared to virtual machines. The virtual eNodeB is converted into a Docker image together with its dependencies and stored in the Docker registry—a repository for virtual images. The virtualized BBU pool is controlled by the OpenStack cloud controller: it can deploy virtual BBUs by instantiating eNodeB Docker containers from the Docker registry and configure the E2E network in less than 60 s.

**CONCLUSIONS**

We present an E2E system, from antenna design and CoMP to FH optimization, BBU pool virtualization, and protocol design, to enhance LTE networks while reducing capital and operational expenditures. The proposal is based on a C-RAN wherein the RRHs are equipped with LCPAAs. It enables dense deployments with CoMP, mitigates ICI thanks to increased DoFs that increase SE and system capacity; and it does so while ensuring cost reductions. Finally, our results provide an indication of the benefits associated with the implementation of the proposed system and can inspire further developments.
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BIOGRAPHIES

MATTEO ARTUSO (mattar@telecom-paris.fr) received a B.S. in information engineering from the University of Padua, Italy, in 2012 and an M.Sc. in telecommunication engineering from the Technical University of Denmark (DTU) in 2014. He is currently the chief technology officer at the software company Unums and a Ph.D. candidate at the Technical University of Denmark, where he is also the Chairman of the local IEEE Student branch. His main research areas are architectures and protocols for next-generation mobile networks, with a special focus on cloud networks, virtualization, and self-organization.

DORA BOVZ [S’15] is a research engineer in the Software Defined Mobile Networks Department at Nokia Bell Labs and a Ph.D. candidate at CentraleSupélec. She holds an engineering degree from ENSEA, France, and a Master’s degree from UCP, France. Her current research is about multi-cell cooperation in cloud-RAN and non-orthogonal multi-user transmissions.

ALEXANDRA CHERRIO is a senior system engineer at MTI Radiocomp. She received her double M.Sc. degree in telecommunication from DTU and Lodz University of Technology, Poland, in 2011. She has worked as a telecommunication trainer conducting courses from ENSEA, France, and a Master’s degree from UCP, France. Her current research is about multi-cell cooperation in cloud-RAN and non-orthogonal multi-user transmissions.

HENRIK L. CHRISTIANSEN is an associate professor in mobile communication at TDU, where he also received his M.Sc.E.E. and Ph.D., specializing in telecommunications. He also has several years of experience in the telecom industry, including startups and mobile phone development. His main areas of research are mobile network architectures, especially mobile backhaul and front haul networks for CRAN and evolution toward 5G mobile.

BRUNO CLERCKX is a senior lecturer at Imperial College London. His area of expertise is communication theory and signal processing for wireless networks. He is the author of 2 books, 120 peer-reviewed international research papers, and 150 standards contributions, and is the inventor of 75 issued or pending patents among which several have been adopted in the specifications of 4G (3GPP LTE/LTE-A) and 5G (5G NR) standards.

LAURA COTTATELLUCI is an assistant professor at EURERCOM. She received her Ph.D. and Master’s degrees from the Technical University of Vienna (2006) and La Sapienza University of Rome (1995), respectively. Prior to her current position she held posts at Telecom Italia, FTW, INRIA, and the University of South Australia. Her research interests lie in the field of communications theory and signal processing for wireless communications, and satellite and complex networks. Her contributions in these fields are based on the application of random matrix theory and game theory.

DAVID GESBERT [F] is a professor and head of the Communications Systems Department of EURERCOM. He also heads the Communications Theory Group. He welcomes collaboration with brilliant students and passionate researchers in the fields of communication theory, signal processing for (wireless) networks, and information theory. He teaches “Advanced Topics in Wireless” (Fall) and “Information Theory” (Fall) at EURERCOM. Recently he was named in the 2014 Thomson-Reuters List of Highly Cited Researchers in Computer Science. He is a member of the Scientific Bureau of the Excellence Laboratory (LABEX) awarded by the French Government to the SophiaTech Campus on “User Centric Networking.” Since 2015, he has held an Advanced ERC grant on the topic of smart device communications.

BOBBY GIZAS is a research scientist and lab engineer at Athens Information Technology (AIT). He received his B.Sc. degree in electrical and electronic engineering in 2002 and his M.Sc. degree in radio frequency and communications engineering in 2004 from the University of Bradford, United Kingdom. Prior to joining AIT, he worked at Alcatel Lucent Bell Labs and Advanced Neuromodulation Systems.

ARAVANATHAN GOPALASINHAM is a research engineer with the Software Defined Mobile Networks Department at Nokia Bell Labs France. He received his double Masters Degree in ICT and Communication Networks from AIT Thailand and TSP France in 2012. His current research focus is on virtualization of next generation mobile networks using SDN/NFV concepts.

FAHREM KHAN [M’02] received his Ph.D. degree in electrical and electronic engineering from Queen’s University Belfast, United Kingdom, in 2012. He is currently a lecturer with the School of Computing and Engineering, University of Huddersfield, United Kingdom. Previously, he worked as a research associate at the Institute for Digital Communications, University of Edinburgh, United Kingdom, where he contributed to project management and research in EU FP7 research projects, HARPP and ADDLE. His research, where he also the field of wireless communications and signal processing with particular focus on cognitive radio, MIMO and millimeter wave communications. Dr. Khan has significant previous teaching and research experience at academic institutions in the UK, Middle East and India.

JEAN-MARC KELIF received his Licence in fundamental physics science from the Louis Pasteur University of Strasbourg, France, and his Engineer degree in material and solid state physical sciences from the University of Paris XII, Villetaneuse, France, in 1984. He specialized in the field of telecommunications and received his Ph.D. degree from the Ecole Nationale Superieure des Telecommunications de Paris (ANST – ParisTech) in 2008. His current research interests include the performance evaluation, modeling, dimensioning, and optimization of wireless telecommunication networks. He collaborates with scientific journals as a reviewer: the Wireless Network Journal, the Journal of Computer and Telecommunications Networking Computer Networks, and IEEE Transactions on Communications. He has been TPC Chair of international conferences: Phsycomnet 2009, IFIP WMNC 2011, and ValueTools 2011.

The proposal is based on a C-RAN wherein the RRHs are equipped with LCPAs. It enables dense deployments with CoMP, it mitigates ICI thanks to increased DoFs that increase SE and system capacity and does so while ensuring cost reductions.
RAFI MÜLLER received his Dipl.-Ing. and Dr.-Ing. degrees with distinction from Friedrich-Alexander University (FAU) Erlangen-Nuremberg, Germany, in 1996 and 1999, respectively. From 2000 to 2004, he directed a research group at the Vienna Telecommunications Research Center, Austria, and taught as an adjunct professor at the Vienna University of Technology. In 2005, he was appointed full professor at the Department of Electronics and Telecommunications at the Norwegian University of Science and Technology (NTNU), Trondheim, Norway. In 2013, he joined the Institute for Digital Communications at FAU Erlangen-Nuremberg. He has held visiting appointments at Princeton University, Institute Eurecom, France, the University of Melbourne, Australia, the University of Oulu, Finland, the National University of Singapore, Babes-Bolyai University, Cluj-Napoca, Romania, Kyoto University, Japan, FAU Erlangen-Nuremberg, Germany, and Munich University of Technology.

BORZOO RASSOULI received his Ph.D. from Imperial College London in 2016. Currently, he is a research associate in the Department of Electrical and Computer Engineering, Department of Electrical and Computer Engineering of Boston University, Massachusetts. Since 2014 he has held a Ph.D. diploma in antennas from Aristotle University of Thessaloniki, Department of Electrical and Computer Engineering.

CONSTANTINOS NTougias is a researcher at AIT and a member of the Broadband Wireless & Sensor Networks (B-WiSE) research group. He is also pursuing a Ph.D. degree jointly at the Department of Electronic Systems, Aalborg University, Denmark, and AIT. The topic of his Ph.D. dissertation is the cross-layer design and optimization for efficient licensed shared access communication with quality of service guarantees.

CONSTANTINOS B. PAPADIAS [S’88, M’96, SM’03, F’13] is the Dean of AIT and head of AIT’s B-WiSE research group. He is also an adjunct professor at Aalborg University. He is a graduate of National Technical University of Athens, Greece (1991), and ENST (Ph.D., 1995). Prior to his current positions he held posts at Eurecom, Stanford University, Bell Labs, Columbia University, and Carnegie Mellon University.

MOHAMMAD ALI SEDAGHAT received his B.Sc degree in electrical engineering from Amirkabir University of Technology, Iran, in 2006, his M.Sc degree in communication systems from Sharif University of Technology, Iran, in 2008, and his Ph.D. degree in wireless communications from NTNU in 2015. Currently, he is a postdoctoral researcher at FAU Erlangen-Nuremberg. His current research interests include the general field of wireless communications with emphasis on MIMO systems for new generations of wireless communications.

THARMALINGAM RATNARAJAH (A96-M05-M05) is currently with the Institute for Digital Communications, University of Edinburgh as a professor in digital communications and signal processing. He is currently the coordinator of the FP7 projects ADEL in the area of licensed shared access for 5G wireless networks. Previously, he was the coordinator of the FP7 project HARPA in the area of highly distributed MIMO, and FP7 Future and Emerging Technologies projects HATUS in the area of interference alignment and CROWN in the area of cognitive radio networks.

LAURENT ROULLET leads the Mobile Networks Virtualization & Verification research group at Nokia Bell Labs, focusing on cloud RAN architectures, fronthauling optimization, software defined networks, and network functions virtualization applied to wireless access. He joined Bell Labs in 2010, and has worked on small cells optimization and interference coordination. He joined Alcatel Lucent Mobile Broadcast in 2006 and led the standardization effort of the DVB-SH (hybrid satellite terrestrial mobile broadcast). From 2001 to 2005 he worked at Udcast, a startup focusing on mobile routers for satellite and DVB networks, and developed the first DVB-H infrastructure in cooperation with Nokia. From 1997 to 2001 he worked at Alcatel Space on onboard processing. He graduated from Ecole Polytechnique and SupAero.

STEPHANE SENECAL received his M.S. in mathematics in 1998 from Joseph Fourier University, and his M.Eng./M.S. in electrical engineering/signal processing in 1999 and Ph.D. in statistical signal processing and telecommunications in 2002, both from Institut National Polytechnique, Grenoble, France. He worked at the Institute of Statistical Mathematics, Tokyo, Japan, and then at France Telecom Japan from 2003 to 2006. In 2007, he joined Orange Labs, Paris, in order to design network deployment and operation optimization methods, and fixed and mobile networks traffic data processing solutions. His research interests span from statistical/machine learning and data mining to statistical modeling and simulation methods.

LIN ZHOU is currently working toward her Ph.D. at the University of Edinburgh. Her research interests include design, modeling, and implementation of compact MIMO transceivers, parasitic reconfigurable arrays, smart antennas, and energy-efficient 5G wireless networks.